

# Performance Analysis of Wearable Microstrip Antennas with Low-Conductivity Materials

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## Introduction

In recent years, with the development of applications in personal communication technologies, wearable antennas have received growing attention [1]. It is anticipated that in the near future, a wide variety of consumer electronics and communication devices will be wearable – built into clothing. Accordingly, textile-based wearable antennas are foreseen as a major component of wearable electronics in emerging personal communication systems.

Conventional wearable antennas are studied and constructed with metallic elements, which can be treated as perfect electric conductors [2], [3]. However, recent advances in material sciences have given rise to electronic devices in which conducting materials in the form of ink are printed on various surfaces [4]. In our research, in order to enhance the flexibility, weight, and conformity of wearable designs, carbon nanostructured conductor inks and conducting polymer inks are proposed and studied as possible conductors in textile-embedded wearable antennas. Since the conductivity of these inks is not high enough for the material to be treated as perfectly conducting, the principal scope of this study is the design, characterization, and development of wearable microstrip antennas with low conductivity materials. In specific, a 2.4-GHz WLAN rectangular microstrip textile antenna design is presented, with low conductivities of the patch and ground materials. Both impedance and radiation properties of the antenna are studied. Numerical results and discussion are based on using two independent commercial electromagnetic tools for antenna modeling and analysis.

## Antenna Model

In this study, CST Microwave Studio<sup>TM</sup> and Ansoft HFSS<sup>TM</sup> are used for the analysis of the antennas. The substrate material is assumed to be lossless, with a dielectric constant of  $\epsilon_r = 2.2$ . As the antenna structure is required to be thin for wearable applications, the substrate thickness is adopted to be 1.5 mm. In addition, due to a delicate nature of conductor ink implementations, it may not be

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appropriate to excite a patch made of conductor polymer by directly connecting the metallic feed to it. Therefore, the patch is fed by a 50- $\Omega$  microstrip transmission line, with a quarter-wave microstrip transformer inserted to match the antenna impedance to 50  $\Omega$ , as shown in Figure 1. The ground conductor is adopted to be three times as large as the patch [5].

The antenna is first designed for the case of all conductors in the structure being made of copper, and the optimum dimensions of the structure, obtained with a series of simulations, are given in Table 1. The conductivity of the patch, transmission lines, and ground is then substantially decreased to model and investigate the performance of the antenna with non-metallic conductors, and to ultimately identify the low-conductivity limit for practical microstrip antenna implementations.

## Results and Discussion

Figure 2 shows the simulated results for the antenna impedance and radiation pattern for conductivities  $\sigma_1 = 5.7 \times 10^7$  S/m (copper),  $\sigma_2 = 5.7 \times 10^5$  S/m, and  $\sigma_3 = 5.7 \times 10^3$  S/m, respectively, of all conductors in the structure. We observe from the figure, that, while the matching properties of the antenna (with respect to 50  $\Omega$ ) appear to be even better for the conductivity value  $\sigma_2$  than for  $\sigma_1$ , they are severely degraded by reducing the conductivity to  $\sigma_3$ . We may say that this level of conductivity makes the antenna impractical in terms of its low received power.

The presented results for the antenna gain, which accounts for conductor losses in the conductors but not for mismatch losses at the port [6], generated by varying  $\theta$  for  $\varphi = 0$  in Figure 1, show similar trends. For copper conductors, the gain pattern is nearly symmetric with respect to the direction defined by  $\theta = 0$ , and has a peak value of 7.3 dB (using both simulation tools). For the conductivity  $\sigma_2$ , both the pattern symmetry and the gain peak value (6 dB) are slightly reduced, but still are acceptable for wearable microstrip antenna applications. However, the antenna gain levels for the conductivity  $\sigma_3$  drop to practically unacceptable values, due to very large conductor losses.

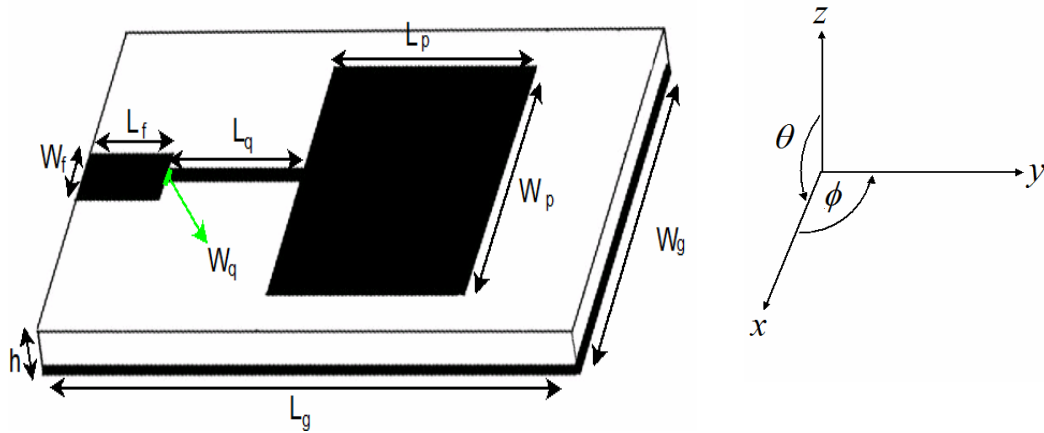
## Conclusions

This paper has presented a 2.4-GHz WLAN rectangular microstrip textile antenna design, along with investigations of the degradation in resonant, matching, and gain properties of the antenna if low conductivity materials are used for all conductors in the structure. The results show that the resonant frequency remains almost the same for lower conductivities, and that the reflection from the antenna port is even smaller for conductivities on the order of  $10^5$  S/m than for copper conductors, the mismatch losses increase to prohibitive levels for conductivities on the order of  $10^3$  S/m. The antenna gain attains reasonable values for  $\sigma \sim 10^5$  S/m. However, for conductors with  $\sigma \sim 10^3$  S/m, the conductor losses dominate the radiated power, resulting in practically unacceptable gain levels. Overall,

conductivity levels of WLAN microstrip textile antennas must be in excess of  $\sim 10^4$  S/m for an acceptable antenna performance in wearable applications.

### References

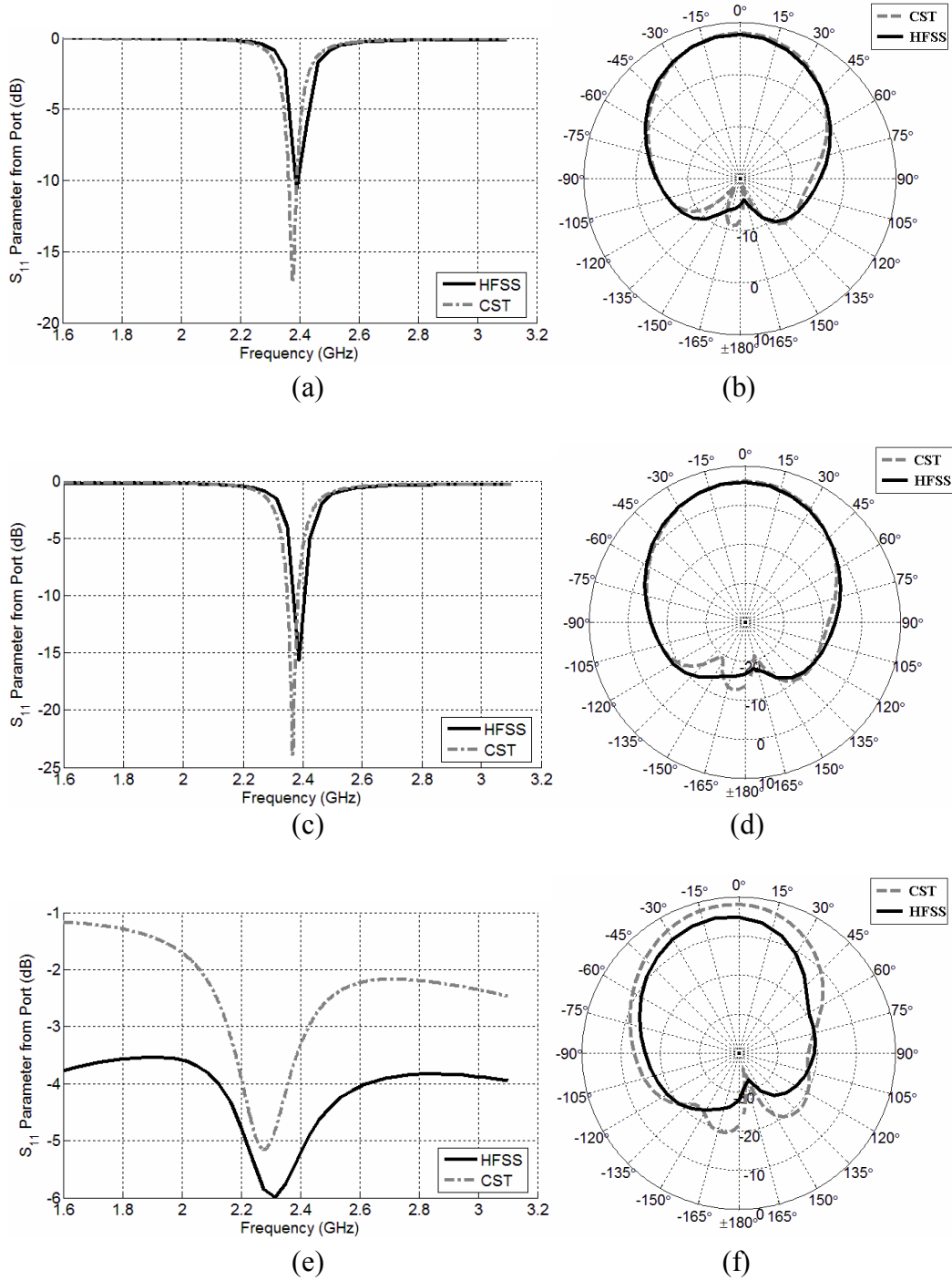
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**Figure 1:** 2.4-GHz WLAN rectangular microstrip textile antenna with low-conductivity materials for all conductors ( $\epsilon_r = 2.2$  for the substrate).

**Table 1:** Dimensions (in mm) of the designed antenna in Figure 1.

| $W_p$ | $L_p$ | $W_q$ | $L_q$ | $W_f$ | $L_f$ | $W_g$ | $L_g$ | $h$ |
|-------|-------|-------|-------|-------|-------|-------|-------|-----|
| 42    | 40.5  | 0.72  | 21.05 | 4.84  | 19.45 | 126   | 121.5 | 1.5 |



**Figure 2:** Simulated results (using CST Microwave Studio<sup>TM</sup> and Ansoft HFSS<sup>TM</sup>) for  $S_{11}$  (reference 50  $\Omega$ ) and gain (in the plane  $\varphi = 0$ ) of the microstrip antenna in Figure 1 (dimensions in Table 1) for conductivities  $\sigma_1 = 5.7 \times 10^7$  S/m (copper) (a)-(b),  $\sigma_2 = 5.7 \times 10^5$  S/m (c)-(d), and  $\sigma_3 = 5.7 \times 10^3$  S/m (e)-(f), of all conductors in the structure.